Complex Gaussian, Continuous Time Stochastic Processes Course: Foundations in Digital Communications

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6th lecture



What did we do last lecture?

- Passband processing is done in complex domain and Gaussian is the most important distribution. Therefore, let's have a look at these distributions:
 - Complex Gaussian and Circular Symmetry (chap 24)
- Most physical signals realizations are continuous. Thus an extension of the notions and a solid understanding of stochastic processes is required.
 - Continuous Time Stochastic Processes (chap 25)

Standard Complex Gaussian

Standard Complex Gaussian

Complex RV *W* whose real and imaginary part are independent $\mathcal{N}(0, 1/2)$ RVs.

$$f_W(w) = \frac{1}{\pi} \mathrm{e}^{-|w|^2}, \quad w \in \mathbb{C}$$

$$\mathbb{E}[W] = 0$$
 and $\operatorname{Var}[W] = \mathbb{E}\left[|W|^2\right] = 1$

Standard complex Gaussian RV are

- proper since $\mathbb{E}[W] = 0$ and $\mathbb{E}[W^2] = 0$
- radially-symmetric since $f_W(w)$ depends on |w| only

Circular Symmetry

Circular Symmetry

CRV *Z* is *circularly-symmetric* if for any $\phi \in [-\pi, \pi)$ we have

$$e^{-i\phi}Z \stackrel{\mathscr{L}}{=} Z$$

If CRV Z has a density, then we have the following equivalences:

- Z is circularly-symmetric
- Z has a radially-symmetric density function
- Z can be written as $Z = Re^{i\Theta}$, RV R and Θ with $R \perp \Theta$, $R \in [0, \infty), \Theta \sim \mathcal{U}([-\pi, \pi))$
- Example: CRV Z = e^{iΦ}, Φ ~ U([−π, π)) is circularly-symmetric, but does not have a density.

Properness and Circular Symmetry

- Circular symmetry considers whole distribution, properness considers first two moments only.
- Every finite-variance circularly-symmetric CRV is proper.
 Not every proper CRV is circularly-symmetric

Proof:

$$\mathbb{E}\left[Z^{k}\right] = e^{-ik\Phi}\mathbb{E}\left[\left(e^{i\Phi}Z\right)^{k}\right] = e^{-ik\Phi}\mathbb{E}\left[Z^{k}\right], \quad \Rightarrow \quad \mathbb{E}\left[Z^{k}\right] = 0, k = 1, 2$$

- Counterexample: CRV *Z* with 1 + i, 1 i, -1 + i, -1 iequiprobable is proper, but $e^{i\pi/4}$ takes values $\sqrt{2}, -\sqrt{2}, \sqrt{2}i, -\sqrt{2}i$, thus not same distribution.
- Result also holds for the random vectors.

Complex Gaussian

Complex Gaussian

A CRV is complex Gaussian if real and imaginary part are jointly real Gaussian. It is centered if the mean is zero.

By some algebraic comparisons we can see

 $\ensuremath{\textit{W}}$ standard complex Gaussian. For a centered complex Gaussian Z

• exists
$$\alpha, \beta \in \mathbb{C}$$
 such that $Z \stackrel{\mathscr{L}}{=} \alpha W + \beta W^*$

2 is proper iff exists $\alpha \in \mathbb{C}$ such that $Z \stackrel{\mathscr{L}}{=} \alpha W$

• For Z proper complex Gaussian from (ii) it follows

$$f_Z(z) = \frac{f_W(z/\alpha)}{|\alpha|^2} = \frac{1}{2|\alpha|^2} e^{-\frac{|z|}{|\alpha|^2}}, z \in \mathbb{C}$$

 \Rightarrow radially-symmetric, which implies circularly-symmetric.

A complex Gaussian is circular-symmetric iff it is proper!

Complex Gaussian Vectors

• Standard complex Gaussian vector: Components are iid standard complex Gaussian:

$$f_W(w) = \frac{1}{\pi^n} \mathrm{e}^{-w^{\dagger}w}, \quad w \in \mathbb{C}^n.$$

- **Circularly-Symmetric**: $e^{-i\phi}Z \stackrel{\mathscr{L}}{=} Z$ for every $\phi \in [-\pi, \pi)$ \Leftrightarrow for every $\alpha \in \mathbb{C}^n$, the CRV $\alpha^T Z$ is circularly-symmetric.
- Z circularly-symmetric \Rightarrow AZ circularly-symmetric, $A \in \mathbb{C}^{n \times m}$
- $\mathbb{E}\left[\mathbf{Z}\mathbf{Z}^{\dagger}\right]$ and $\mathbb{E}\left[\mathbf{Z}\mathbf{Z}^{T}\right]$ specify a centered complex Gaussian vector \mathbf{Z}

 $\Rightarrow \mathbb{E}\left[\mathbf{Z}\mathbf{Z}^{\dagger}\right]$ specifies a *proper* complex Gaussian vector

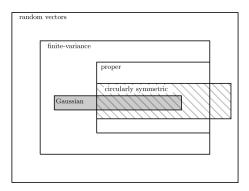
- Complex Gaussian Vector: $\exists A, B \in \mathbb{C}^{n \times m}$ and $\mu \in \mathbb{C}^{n}$ such that $Z \stackrel{\mathscr{L}}{=} AW + BW^* + \mu$, $W \in \mathbb{C}^{m}$ standard complex Gaussian
- Complex Gaussian vector is proper iff it is circularly-symmetric

Proper Complex Gaussian Vectors

• Pdf of **proper** complex Gaussian vector Z with K > 0

$$f_{\mathbf{Z}}(z) = \frac{1}{\pi^n \det K} e^{-z^{\dagger} K^{-1} z}, \quad z \in \mathbb{C}^n$$

• Relationship between complex random vectors:



Let's take a break!

Continuous-Time Stochastic Processes (SP)

• SP $(X(t), t \in \mathbb{R})$ defined on common probability space (Ω, \mathcal{F}, P)

 $X: \Omega \times \mathbb{R} \to \mathbb{R}, \qquad (\omega, t) \mapsto X(\omega, t)$

- Random variable: $\omega \mapsto X(\omega, t)$
- Function of time: $t \mapsto X(\omega, t)$ (trajectory)
- SP (X(t)) is of zero mean (finite variance) if RV X(t) is centered (finite variance) for every t ∈ ℝ

Finite dimensional distributions (FDD) of a SP

Family of joint distributions $(X(t_1), \ldots, X(t_n))$ for any positive integer n and epochs $t_1, \ldots, t_n \in \mathbb{R}$

- (X(t)) and (Y(t)) are **independent SP** if $(X(t_1), \ldots, X(t_n)$ and $(Y(t_1), \ldots, Y(t_n))$ are independent $\forall n$ and $t_1, \ldots, t_n \in \mathbb{R}$.
- CDF: $F_n(\xi_1, \ldots, \xi_n; t_1, \ldots, t_n) \triangleq \mathbb{P}[X(t_1) \leq \xi_1, \ldots, X(t_n) \leq \xi_n]$

Kolmogorov's Existence Theorem

Kolmogorov's Existence Theorem

Let $(G_n)_n$ sequence of functions $G_n : \mathbb{R}^n \times \mathbb{R}^n \to [0, 1]$ satisfying

- $G_n(\cdot; t_1, \ldots, t_n)$ is a valid joint distribution $\forall n \ge 1$ and $t_1, \ldots, t_n \in \mathbb{R}$
- symmetry property holds for all n, t_k, ξ_k and permutations π $G_n(\xi_{\pi(1)}, \ldots, \xi_{\pi(n)}; t_{\pi(1)}, \ldots, t_{\pi(n)}) = G_n(\xi_1, \ldots, \xi_n; t_1, \ldots, t_n)$
- and **consistency property** for all t_k , ξ_k $\lim_{\xi_n \to \infty} G_n(\xi_1, \dots, \xi_n; t_1, \dots, t_n) = G_{n-1}(\xi_1, \dots, \xi_{n-1}; t_1, \dots, t_{n-1})$

then \exists SP (*X*(*t*)) whose FDDs are given by {*G_n*(·; ·)} in the sense

$$\mathbb{P}\left[X(t_1) \leq \xi_1, \dots, X(t_n) \leq \xi_n\right] = G_n(\xi_1, \dots, \xi_n; t_1, \dots, t_n)$$

Symmetry and consistency are sufficient for existence of FDDs of some SP

Gaussian Stochastic Process (SP)

Gaussian stochastic process

SP (*X*(*t*)) is a **Gaussian stochastic process** if $(X(t_1), ..., X(t_n))^T$ is a Gaussian random vector for every *n*.

One reason why Gaussian SP are tractable:

FDDs for *centered* Gaussian SP: All its FDDs are determined by autocovariance function $(t_1, t_2) \mapsto \text{Cov} [X(t_1), X(t_2)]$

Proof: Since Gaussian SP is centered, mean is zero and covariance matrix for any *n*-vector is given by $(Cov[X(t_j), X(t_k)])_{1 \le j,k \le n}$.

• $\tau \mapsto \text{Cov} [X(t), X(t + \tau)]$ if wide-sense stationary Gaussian SP

Stationary Continuous-Time Process

Stationary

A SP is stationary (aka strict sense stationary, strongly stationary) if

$$(X(t_1 + \tau), \dots, X(t_n + \tau)) \stackrel{\mathscr{L}}{=} (X(t_1), \dots, X(t_n))$$

Wide-Sense Stationary

A SP is **wide-sense stationary** (aka second-order stationary, weakly stationary) if the following are met

- it is of finite variance,
- 2 $\mathbb{E}[X(t)] = \mathbb{E}[X(t+\tau)]$ for all $t, \tau \in \mathbb{R}$,
- covariance between samples

 $\operatorname{Cov} \left[X(t_1) X(t_2) \right] = \operatorname{Cov} \left[X(t_1 + \tau) X(t_2 + \tau) \right] \quad \forall t_1, t_2, \tau \in \mathbb{R}.$

• Autocovariance fct of WSS SP $K_{XX}(\tau) \triangleq \text{Cov} [X(t + \tau)X(t)]$

Stationary Gaussian Stochastic Process

- Stationary Gaussian Stochastic Processes
 - A Gaussian SP is stationary iff it is WSS.
 - FDDs of a centered stationary Gaussian SP are fully specified by its autocovariance function.
- Relationship between stochastic processes

stochas	finite-variance	
	WSS (strictly) stationary Gaussian	

Properties Autocovariance Function

- If the autocovariance fct of WSS SP is continuous at the origin, then it is a uniformly continuous function.
- The autocovariance function $K_{XX}(\tau)$ of a WSS SP X(t) is a
 - symmetric function: $K_{XX}(\tau) = K_{XX}(-\tau)$
 - positive definite function: $\sum_{\nu=1}^{n} \sum_{\nu'=1}^{n} \alpha_{\nu} \alpha_{\nu'} K_{XX}(t_{\nu} t_{\nu'}) \ge 0$
- Proof: 5-minute exercise.

Every symmetric positive definite function is the autocovariance function of some stationary Gaussian SP.

Proof: Construct covariance matrix of centered Gaussian vector and then use Kolmogorov's Existence Theorem.

PSD of a Continuous-Time SP

PSD

WSS SP (X(t)) is of PSD S_{XX} if S_{XX} is a non-negative, symmetric, integrable function with

$$K_{XX}(\tau) = \int_{-\infty}^{\infty} S_{XX}(f) \mathrm{e}^{i2\pi f\tau} \mathrm{d}f, \quad \tau \in \mathbb{R}.$$

$$\operatorname{Var}[X(t)] = K_{XX}(0) = \int_{-\infty}^{\infty} S_{XX}(f) \, \mathrm{d}f$$

- Every non-negative, symmetric, integrable fct is the PSD of some stationary Gaussian SP whose autocovariance function is continuous.
- A more general concept: Not every WSS SP with a continuous acf has a PSD, but it has a *spectral distribution function*.

Average Function

 A SP is *measurable* if the mapping (ω, t) → X(ω, t) is a measurable mapping.

Power in Centered WSS SP

(X(t)) a measurable, centered, WSS SP with acf K_{XX} , then the RV

$$\omega \mapsto \frac{1}{b-a} \int_{a}^{b} X^{2}(\omega, t) \mathrm{d}t$$

satisfies $\frac{1}{b-a}\mathbb{E}\left[\int_{a}^{b} X^{2}(\omega, t) dt\right] = K_{XX}(0)$, which denotes the power.

Proof:

$$\mathbb{E}\left[\int_{a}^{b} X^{2}(\omega, t) \mathrm{d}t\right] = \int_{a}^{b} \mathbb{E}\left[X^{2}(\omega, t)\right] \mathrm{d}t = \int_{a}^{b} K_{XX}(0) \mathrm{d}t = (b-a)K_{XX}(0)$$

Integral of a SP – Mean Value

• For detecting and processing of continuous-time signals we are interested in stochastic integrals.

$$\omega\mapsto\int_{-\infty}^{\infty}X(\omega,t)s(t)\mathrm{d}t$$

 $s : \mathbb{R} \to \mathbb{R}$ integrable function, (X(t)) measurable WSS SP

- We derive heuristically mean and variance for centered WSS SP, the extension to non-centered X(ω, t) + μ is straightforward.
 - Mathematically, some issues have to be resolved: $t \mapsto X(\omega, t)s(t)$ has to be integrable for almost all ω . The result of integration has to be a RV (check book and references).
- Mean:

$$\mathbb{E}\left[\int_{-\infty}^{\infty} X(\omega, t)s(t)dt\right] = \int_{-\infty}^{\infty} \mathbb{E}\left[X(\omega, t)\right]s(t)dt = \mathbb{E}\left[X(0)\right]\int_{-\infty}^{\infty} s(t)dt$$

Integral of a SP – Variance

• Variance:

$$\operatorname{Var}\left[\int_{-\infty}^{\infty} X(\omega, t)s(t)dt\right] = \mathbb{E}\left[\left(\int_{-\infty}^{\infty} X(\omega, t)s(t)dt\right)^{2}\right]$$
$$= \mathbb{E}\left[\int_{-\infty}^{\infty}\int_{-\infty}^{\infty} X(\omega, t)s(t)X(\omega, \tau)s(\tau)dtd\tau\right]$$
$$= \int_{-\infty}^{\infty}\int_{-\infty}^{\infty} s(t)\mathbb{E}\left[X(\omega, t)X(\omega, \tau)\right]s(\tau)dtd\tau$$
$$= \int_{-\infty}^{\infty} K_{XX}(\sigma)\underbrace{\int_{-\infty}^{\infty} s(\sigma + \tau)s(\tau)d\tau}_{=R_{ss}(\sigma)}d\sigma = \int_{-\infty}^{\infty} K_{XX}(\sigma)R_{ss}(\sigma)d\sigma$$

• Frequency domain: $\operatorname{Var}\left[\int_{-\infty}^{\infty} X(\omega, t)s(t)dt\right] = \int_{-\infty}^{\infty} S_{XX}(f)|\hat{s}(f)|^2 df$

Linear Functionals of Stationary Gaussian Processes

 The extension of the argument that the linear combination of Gaussian RV is a Gaussian RV extends for the linear combination of Gaussian WSS SP (very valuable result)!

Additionally to the previous let (*X*(*t*)) be **Gaussian SP**, $\alpha_{\nu} \in \mathbb{R}$, then

$$\omega \mapsto \int_{-\infty}^{\infty} X(\omega, t) s(t) dt + \sum_{\nu=1}^{n} \alpha_{\nu} X(\omega, t_{\nu})$$

almost always exists and result is a **Gaussian RV** with mean $\mathbb{E}[X(0)]\left(\int_{-\infty}^{\infty} s(t)dt + \sum_{\nu=1}^{n} \alpha_{\nu}\right)$ and variance

$$\operatorname{Var}\left[\int_{-\infty}^{\infty} X(\omega,t)s(t)dt + \sum_{\nu=1}^{n} \alpha_{\nu}X(\omega,t_{\nu})\right] = \int_{-\infty}^{\infty} K_{XX}(\sigma)R_{ss}(\sigma)d\sigma$$
$$+ \sum_{\nu=1}^{n} \sum_{\nu'=1}^{n} \alpha_{\nu}\alpha_{\nu'}K_{XX}(t_{\nu}-t_{\nu'}) + 2\sum_{\nu=1}^{n} \alpha_{\nu} \int_{-\infty}^{\infty} s(t)K_{XX}(t-t_{\nu})dt$$

Joint Distribution of Linear Functionals

• Even more: If (*X*(*t*)) is a Gaussian SP, then the joint distribution of a collection of linear functionals is jointly Gaussian!

Let $s_j : \mathbb{R} \to \mathbb{R}$ measurable functions, $\alpha_{j,\nu} \in \mathbb{R}$. Then the linear functionals (RVs)

$$\omega \mapsto \int_{-\infty}^{\infty} X(\omega, t) s_j(t) dt + \sum_{\nu=1}^{n_j} \alpha_{j,\nu} X(\omega, t_{j,\nu}), \quad j = 1, \dots, m$$

of a measurable, Gaussian WSS SP (X(t)) are jointly Gaussian.

- Generally, second-order properties of linear functionals of measurable WSS SP can be characterized by the acf K_{XX}.
 - In the Gaussian case, the mean and covariance matrix completely specifies the joint distribution.

Filtering of WSS Processes $\int_{-\infty}^{\infty} X(\sigma)h(t-\sigma)d\sigma$

$$\begin{pmatrix} X(t) \end{pmatrix} \qquad \qquad \begin{pmatrix} X(t) \end{pmatrix} \star \mathbf{h} \\ & & \end{pmatrix}$$

(i) Passing a WSS SP through a stable filter produces a WSS SP with acf K_{YY} = K_{XX} ★ R_{hh}, and E [X(t)Y(t + τ)] = (K_{XX} ★ h)(τ)
 (X(t)) and (Y(t)) are said *jointly wide-sense stationary*.

(ii) An input with PSD S_{XX} , then output PSD $S_{YY}(f) = S_{XX}(f)|\hat{h}(f)|^2$ (iii) If the input is Gaussian, then so is the output.

• Additionally, (X(t)) and (Y(t)) are jointly Gaussian.

If *h* satisfies ∫[∞]_{-∞} h²(t)(1 + t²)dt < ∞, then the convolution is defined for all *t* and almost all ω (universality).

Wiener-Khinchin Theorem

Wiener-Khinchin Theorem

Let (X(t)) a measurable, centered, WSS SP with acf K_{XX} is passed through a stable filter with impulse response h, then the average power is

Power of
$$X \star h = (K_{XX} \star R_{hh})(0) = \int_{-\infty}^{\infty} K_{XX}(\tau) R_{hh}(\tau) d\tau$$

If additionally (X(t)) is of PSD S_{XX} , then

Power of
$$X \star h = \int_{-\infty}^{\infty} S_{XX}(f) |\hat{h}(f)|^2 df$$

 This result can be used to show that PSD and operational PSD of a WSS SP are almost the same and both exist if one exits.

White Gaussian Noise

• Most important continuous-time SP, slightly differently defined:

White Gaussian Noise

(N(t)) is white Gaussian noise of double-sided spectral density $N_0/2$ with respect to bandwidth W if (N(t)) is measurable, stationary, centered, Gaussian SP with PSD satisfying

$$S_{NN}(f) = N_0/2 \qquad f \in [-W, W]$$

• Using the previous, many key properties can be derived, e.g.

- s(t) integrable W-bandlimited fct $\int_{-\infty}^{\infty} N(t)s(t)dt \sim \mathcal{N}(0, N_0/2||s||^2)$
- *m* such functions lead to jointly Gaussian RVs
- If they are orthonormal, then the RVs are iid $\mathcal{N}(0, N_0/2)$
- $K_{NN} \star s = N_0/2s$ and $\operatorname{Cov}\left[\int_{-\infty}^{\infty} N(\sigma)s(\sigma)d\sigma, N(t)\right] = N_0/2s(t)$
- **Extension:** White noise in passband $(f \in [f_0 W, f_0 + W])!$

Outlook - Assignment

- Complex Gaussian and Circular Symmetry
- Continuous Time Stochastic Processes

Next lecture

Detection in White Noise; Non-coherent Detection and Nuisance Parameters

- Reading assignment: Chap 26-27
- Homework:
 - Problems in textbook: Exercises 24.1, 24.3, 24.4, 24.8, 25.1, 25.5, 25.10, and 25.11
 - Deadline: Dec 11